

NCV4276, NCV4276A

400 mA Low-Drop Voltage Regulator

The NCV4276 is a 400 mA output current integrated low dropout regulator family designed for use in harsh automotive environments. It includes wide operating temperature and input voltage ranges. The device is offered with fixed output voltage options of 1.8 V and 2.5 V with 4% output voltage accuracy while the 3.3 V, 5.0 V, and adjustable voltage versions are available either in 2% or 4% output voltage accuracy. It has a high peak input voltage tolerance and reverse input voltage protection. It also provides overcurrent protection, overtemperature protection and inhibit for control of the state of the output voltage. The NCV4276 family is available in DPAK and D²PAK surface mount packages. The output is stable over a wide output capacitance and ESR range.

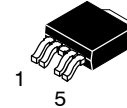
Features

- 2.5 V and 1.8 V $\pm 4\%$ Output Voltage
- 3.3 V, 5.0 V, and Adjustable Voltage Version (from 2.5 V to 20 V) $\pm 4\%$ or $\pm 2\%$ Output Voltage
- 400 mA Output Current
- 500 mV (max) Dropout Voltage (5.0 V Output)
- Inhibit Input
- Very Low Current Consumption
- Fault Protection
 - ◆ +45 V Peak Transient Voltage
 - ◆ -42 V Reverse Voltage
 - ◆ Short Circuit
 - ◆ Thermal Overload
- NCV Prefix for Automotive and Other Applications Requiring Site and Control Changes
- Pb-Free Packages are Available

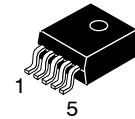


ON Semiconductor®

<http://onsemi.com>



**DPAK
5-PIN
DT SUFFIX
CASE 175AA**



**D²PAK
5-PIN
DS SUFFIX
CASE 936A**

DEVICE MARKING INFORMATION

See general marking information in the device marking section on page 20 of this data sheet.

ORDERING INFORMATION

See detailed ordering and shipping information in the ordering information section on page 21 of this data sheet.

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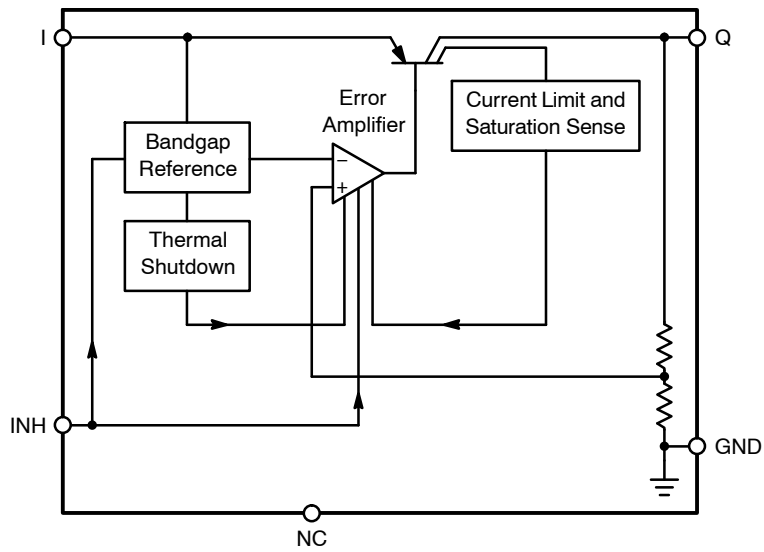


Figure 1. 4276 Block Diagram

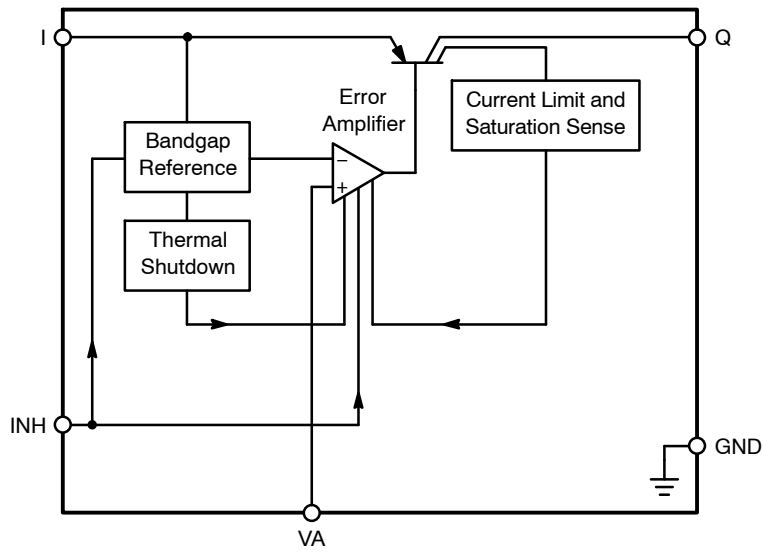


Figure 2. 4276 Adjustable Block Diagram

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PIN FUNCTION DESCRIPTION

Pin No.	Symbol	Description
1	I	Input; Battery Supply Input Voltage.
2	INH	Inhibit; Set low-to inhibit.
3	GND	Ground; Pin 3 internally connected to heatsink.
4	NC / VA	Not connected for fixed voltage version / Voltage Adjust Input for adjustable voltage version; use an external voltage divider to set the output voltage
5	Q	Use 22 μ F, ESR < 2.5 Ω at 10 kHz to ground with the 5.0 V and adjustable regulators. See Figures 3, 4, and 5. Use 10 μ F, ESR < 1.8 Ω at 10 kHz to ground with the 3.3 V, 2.5 V, and 1.8 V regulators. See Figures 3 and 6.

MAXIMUM RATINGS*

Rating	Symbol	Min	Max	Unit
Input Voltage	V_I	-42	45	V
Input Peak Transient Voltage	V_I	-	45	V
Inhibit INH Voltage	V_{INH}	-42	45	V
Output Voltage	V_Q	-1.0	40	V
Ground Current	I_q	-	100	mA
Input Voltage Operating Range	V_I	$V_Q + 0.5$ V or 4.5 V (Note 1)	40	V
ESD Susceptibility	(Human Body Model)	-	4.5	kV
	(Machine Model)	-	250	V
	(Charged Device Model)	-	1.25	kV
Junction Temperature	T_J	-40	150	$^{\circ}$ C
Storage Temperature	T_{stg}	-50	150	$^{\circ}$ C

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

*During the voltage range which exceeds the maximum tested voltage of I, operation is assured, but not specified. Wider limits may apply. Thermal dissipation must be observed closely.

LEAD TEMPERATURE SOLDERING REFLOW (Note 2)

Lead Temperature Soldering	T_{SLD}			$^{\circ}$ C
Reflow (SMD styles only), Leaded, 60–150 s above 183, 30 s max at peak		-	240	
Reflow (SMD styles only), Lead Free, 60–150 s above 217, 40 s max at peak		-	265	
Wave Solder (through hole styles only), 12 sec max		-	310	

THERMAL CHARACTERISTICS

Characteristic	Test Conditions (Typical Value)	Unit
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DPAK 5-PIN PACKAGE

	Min Pad Board (Note 3)	1" Pad Board (Note 4)	
Junction-to-Tab (ψ_{JLx} , ψ_{JLx})	4.2	4.7	C/W
Junction-to-Ambient ($R_{\theta JA}$, θ_{JA})	100.9	46.8	C/W

D²PAK 5-PIN PACKAGE

	0.4 sq. in. Spreader Board (Note 5)	1.2 sq. in. Spreader Board (Note 6)	
Junction-to-Tab (ψ_{JLx} , ψ_{JLx})	3.8	4.0	C/W
Junction-to-Ambient ($R_{\theta JA}$, θ_{JA})	74.8	41.6	C/W

1. Minimum $V_I = 4.5$ V or ($V_Q + 0.5$ V), whichever is higher.
2. Per IPC / JEDEC J-STD-020C.
3. 1 oz. copper, 0.26 inch² (168 mm²) copper area, 0.062" thick FR4.
4. 1 oz. copper, 1.14 inch² (736 mm²) copper area, 0.062" thick FR4.
5. 1 oz. copper, 0.373 inch² (241 mm²) copper area, 0.062" thick FR4.
6. 1 oz. copper, 1.222 inch² (788 mm²) copper area, 0.062" thick FR4.

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ELECTRICAL CHARACTERISTICS ($V_I = 13.5\text{ V}$; $-40^\circ\text{C} < T_J < 150^\circ\text{C}$; unless otherwise noted.)

Characteristic	Symbol	Test Conditions	NCV4276			NCV4276A			Unit
			Min	Typ	Max	Min	Typ	Max	
OUTPUT									
Output Voltage, 5.0 V Version	V_Q	$5.0\text{ mA} < I_Q < 400\text{ mA}$, $6.0\text{ V} < V_I < 28\text{ V}$	4.8	5.0	5.2	4.9	5.0	5.1	V
Output Voltage, 5.0 V Version	V_Q	$5.0\text{ mA} < I_Q < 200\text{ mA}$, $6.0\text{ V} < V_I < 40\text{ V}$	4.8	5.0	5.2	4.9	5.0	5.1	V
Output Voltage, 3.3 V Version	V_Q	$5.0\text{ mA} < I_Q < 400\text{ mA}$, $4.5\text{ V} < V_I < 28\text{ V}$	3.168	3.3	3.432	3.234	3.3	3.366	V
Output Voltage, 3.3 V Version	V_Q	$5.0\text{ mA} < I_Q < 200\text{ mA}$, $4.5\text{ V} < V_I < 40\text{ V}$	3.168	3.3	3.432	3.234	3.3	3.366	V
Output Voltage, 2.5 V Version	V_Q	$5.0\text{ mA} < I_Q < 400\text{ mA}$, $4.5\text{ V} < V_I < 28\text{ V}$	2.4	2.5	2.6	–	–	–	V
Output Voltage, 2.5 V Version	V_Q	$5.0\text{ mA} < I_Q < 200\text{ mA}$, $4.5\text{ V} < V_I < 40\text{ V}$	2.4	2.5	2.6	–	–	–	V
Output Voltage, 1.8 V Version	V_Q	$5.0\text{ mA} < I_Q < 400\text{ mA}$, $4.5\text{ V} < V_I < 28\text{ V}$	1.728	1.8	1.872	–	–	–	V
Output Voltage, 1.8 V Version	V_Q	$5.0\text{ mA} < I_Q < 200\text{ mA}$, $4.5\text{ V} < V_I < 40\text{ V}$	1.728	1.8	1.872	–	–	–	V
Output Voltage, Adjustable Version	AV_Q	$5.0\text{ mA} < I_Q < 400\text{ mA}$ $V_{Q+1} < V_I < 40\text{ V}$ $V_I > 4.5\text{ V}$	–4%	–	+4%	–2%	–	+2%	V
Output Current Limitation	I_Q	$V_Q = 90\% V_{Q\text{TYP}}$ ($V_{Q\text{TYP}} = 2.5\text{ V}$ for ADJ version)	400	700	1100	400	700	1100	mA
Quiescent Current (Sleep Mode) $I_q = I_I - I_Q$	I_q	$V_{\text{INH}} = 0\text{ V}$	–	–	10	–	–	10	μA
Quiescent Current, $I_q = I_I - I_Q$	I_q	$I_Q = 1.0\text{ mA}$	–	130	220	–	130	200	μA
Quiescent Current, $I_q = I_I - I_Q$	I_q	$I_Q = 250\text{ mA}$	–	10	15	–	10	15	mA
Quiescent Current, $I_q = I_I - I_Q$	I_q	$I_Q = 400\text{ mA}$	–	25	35	–	25	35	mA
Dropout Voltage, 5.0 V Version 3.3 V Version 2.5 V Version 1.8 V Version Adjustable Version	V_{DR}	$I_Q = 250\text{ mA}$, $V_{\text{DR}} = V_I - V_Q$ $V_I = 5.0\text{ V}$ $V_I = 4.5\text{ V}$ $V_I = 4.5\text{ V}$ $V_I = 4.5\text{ V}$ $V_I > 4.5\text{ V}$	–	250	500	–	–	–	mV V V V mV
Dropout Voltage (5.0 V Version)	V_{DR}	$I_Q = 250\text{ mA}$ (Note 7)	–	–	–	–	250	500	mV
Load Regulation	$\Delta V_{Q,\text{LO}}$	$I_Q = 5.0\text{ mA}$ to 400 mA	–	10	35	–	3.0	20	mV
Line Regulation	ΔV_Q	$\Delta V_I = 12\text{ V}$ to 32 V , $I_Q = 5.0\text{ mA}$	–	2.5	25	–	4.0	15	mV
Power Supply Ripple Rejection	PSRR	$f_r = 100\text{ Hz}$, $V_r = 0.5\text{ V}_{\text{PP}}$	–	60	–	–	70	–	dB
Temperature Output Voltage Drift	dV_Q/dT	–	–	0.5	–	–	0.5	–	mV/K

INHIBIT

Inhibit Voltage, Output High	V_{INH}	$V_Q \geq V_{Q\text{MIN}}$	–	2.8	3.5	–	2.3	2.8	V
Inhibit Voltage, Output Low (Off)	V_{INH}	$V_Q \leq 0.1\text{ V}$	0.5	1.7	–	1.8	2.2	–	V
Input Current	I_{INH}	$V_{\text{INH}} = 5.0\text{ V}$	5.0	10	20	5.0	10	20	μA

THERMAL SHUTDOWN

Thermal Shutdown Temperature*	T_{SD}	$I_Q = 5.0\text{ mA}$	150	–	210	150	–	210	$^\circ\text{C}$
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*Guaranteed by design, not tested in production.

7. Measured when the output voltage V_Q has dropped 100 mV from the nominal valued obtained at $V = 13.5\text{ V}$.

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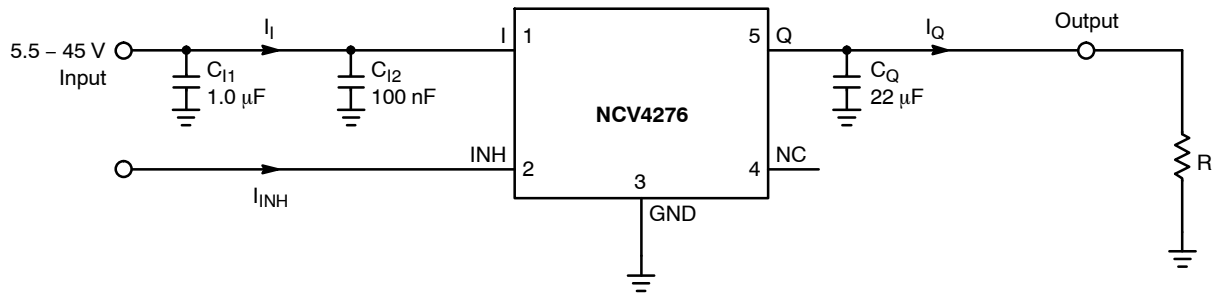


Figure 3. Applications Circuit; Fixed Voltage Version

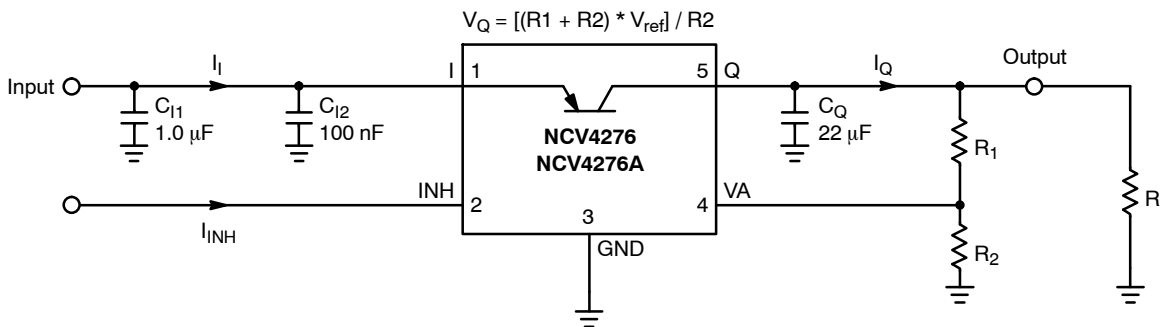


Figure 4. Applications Circuit; Adjustable Voltage Version

TYPICAL PERFORMANCE CHARACTERISTICS

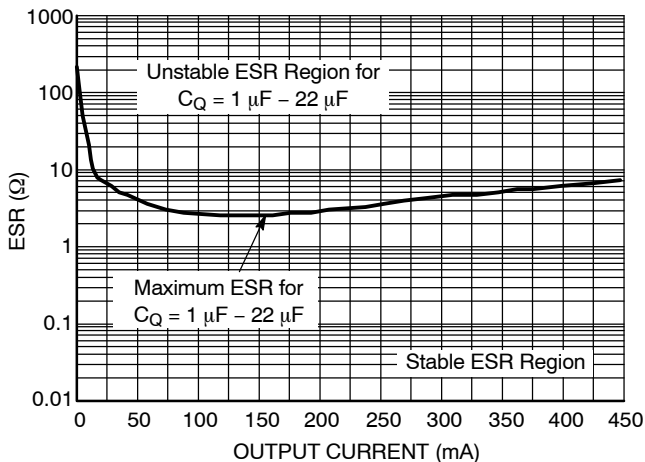


Figure 5. Output Stability with Output Capacitor ESR, 5.0 V and Adjustable Regulator

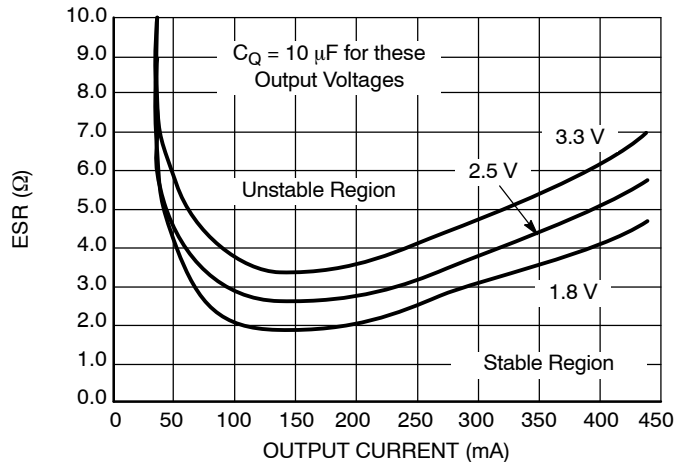


Figure 6. Output Stability with Output Capacitor ESR, 1.8 V, 2.5 V, 3.3 V Regulators

TYPICAL PERFORMANCE CHARACTERISTICS – 4276 Version

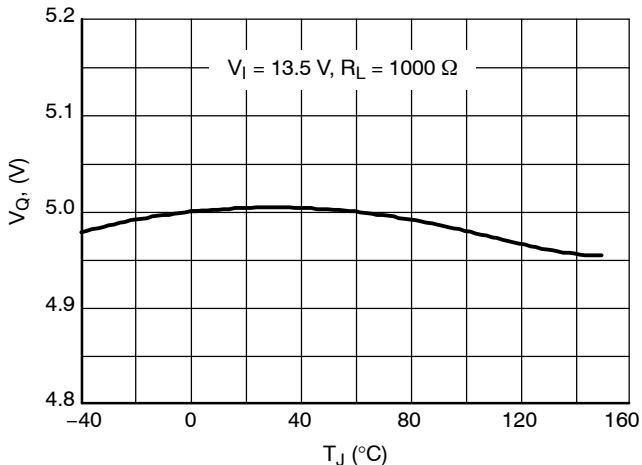


Figure 7. Output Voltage V_Q vs. Temperature T_J , 5.0 V Version

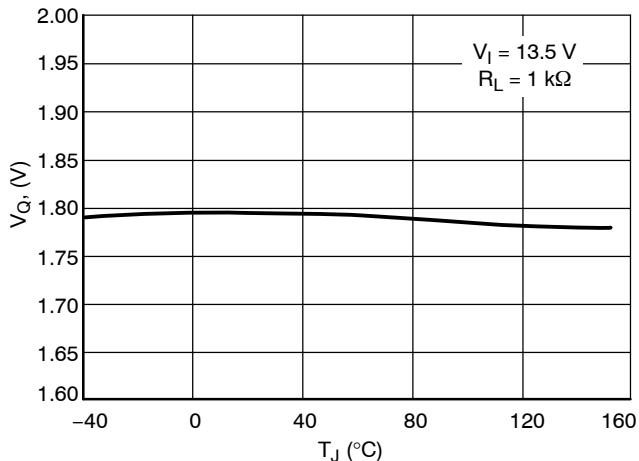


Figure 8. Output Voltage V_Q vs. Junction Temperature T_J , 1.8 V Version

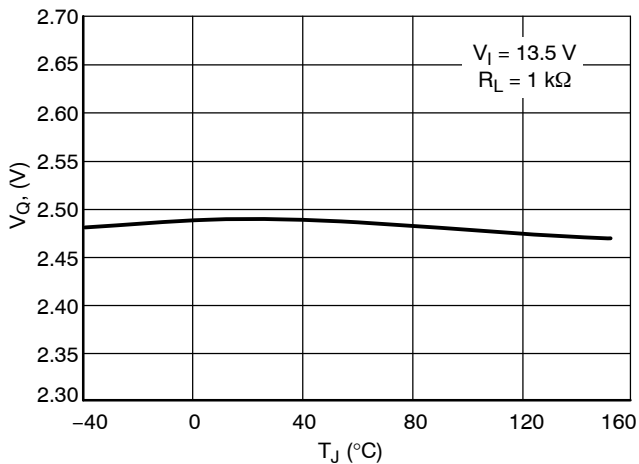


Figure 9. Output Voltage V_Q vs. Junction Temperature T_J , 2.5 V Version

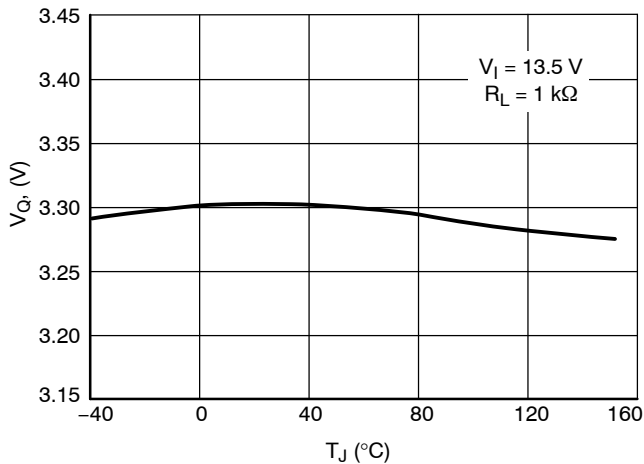


Figure 10. Output Voltage V_Q vs. Junction Temperature T_J , 3.3 V Version

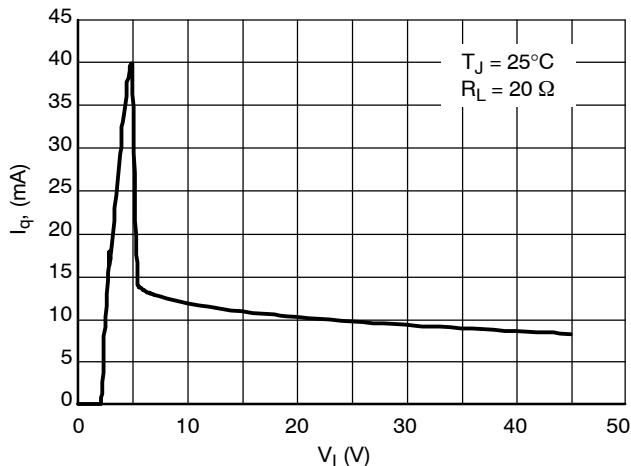


Figure 11. Current Consumption I_q vs. Input Voltage V_I , 5.0 V Version

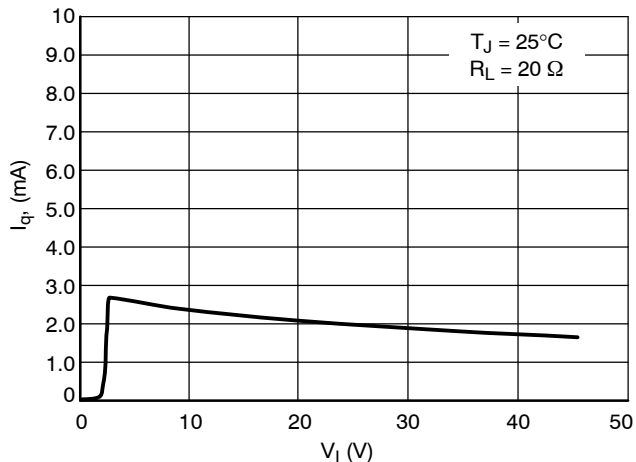


Figure 12. Current Consumption I_q vs. Input Voltage V_I , 1.8 V Version

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TYPICAL PERFORMANCE CHARACTERISTICS – 4276 Version

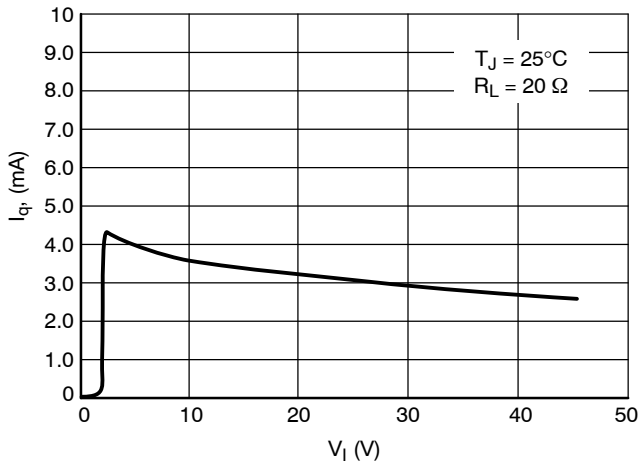


Figure 13. Current Consumption I_q vs. Input Voltage V_i , 2.5 V Version

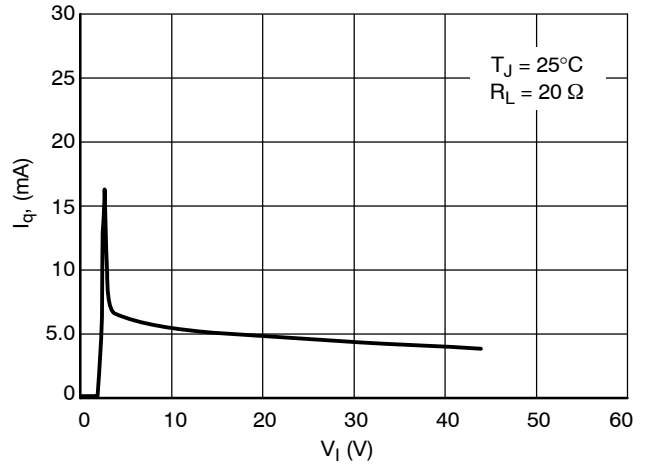


Figure 14. Current Consumption I_q vs. Input Voltage V_i , 3.3 V Version

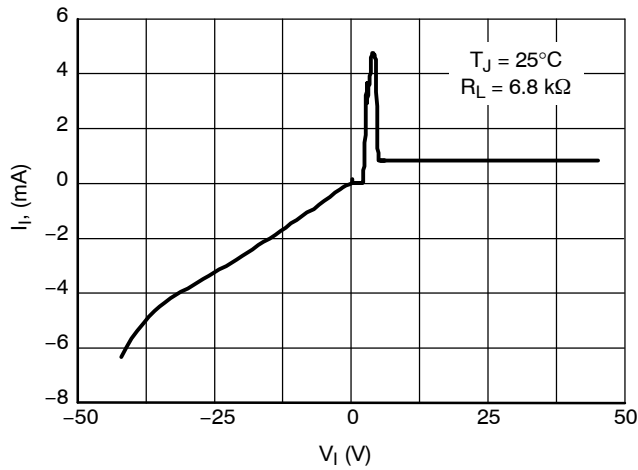


Figure 15. High Voltage Behavior

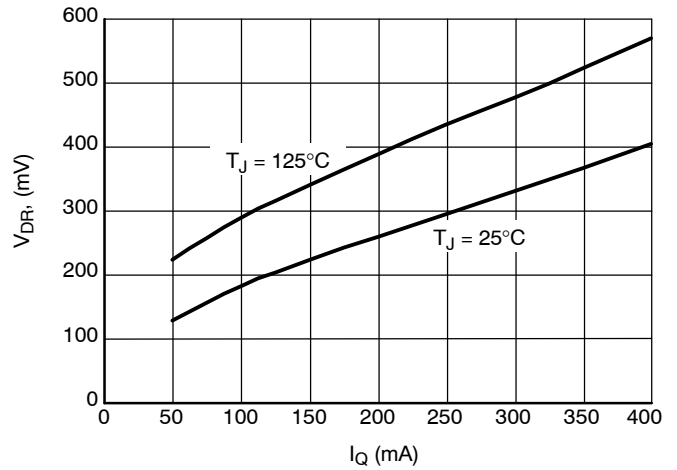


Figure 16. Dropout Voltage V_{DR} vs. Output Current I_Q , 5.0 V Version

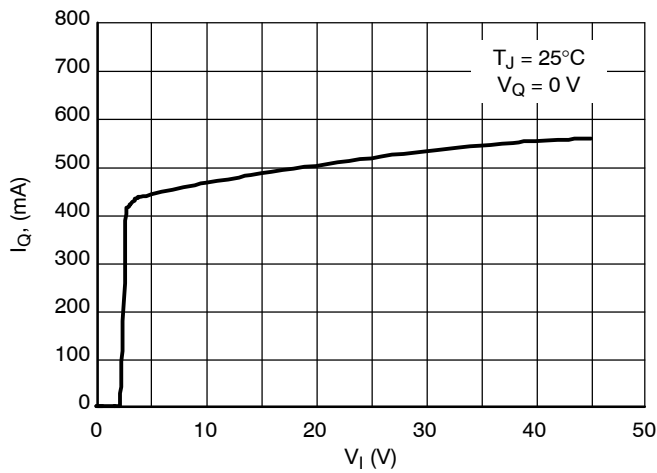


Figure 17. Maximum Output Current I_Q vs. Input Voltage V_i

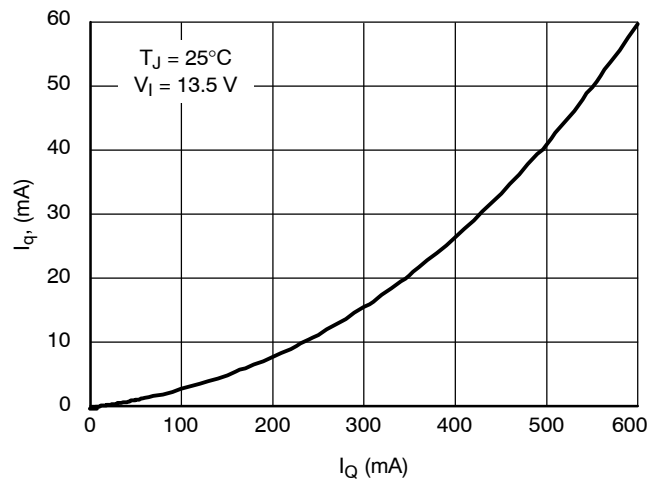


Figure 18. Current Consumption I_q vs. Output Current I_Q (High Load)

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TYPICAL PERFORMANCE CHARACTERISTICS – 4276 Version

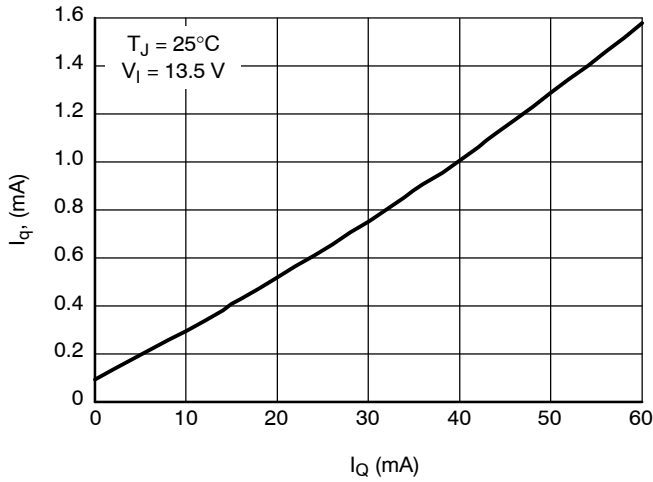


Figure 19. Current Consumption I_Q vs. Output Current I_Q (Low Load)

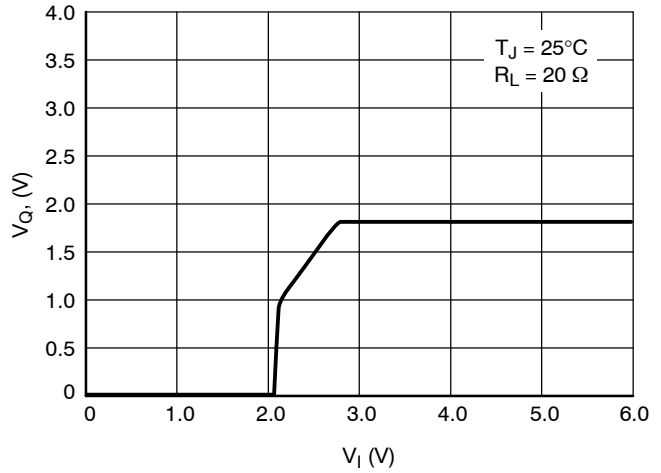


Figure 20. Output Voltage V_Q vs. Input Voltage V_I , 1.8 V Version

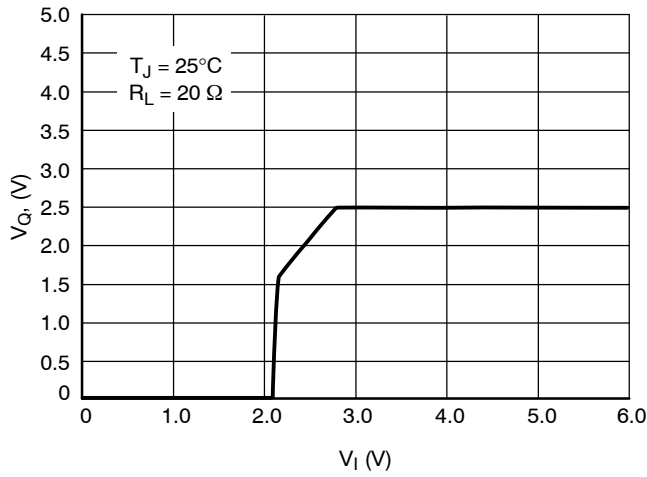


Figure 21. Output Voltage V_Q vs. Input Voltage V_I , 2.5 V Version

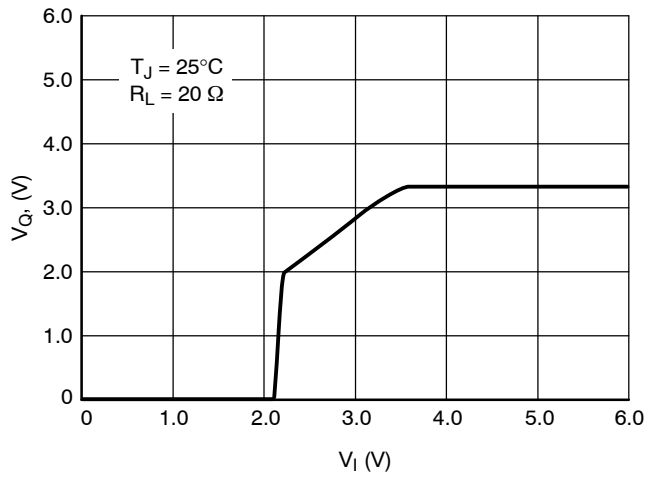


Figure 22. Output Voltage V_Q vs. Input Voltage V_I , 3.3 V Version

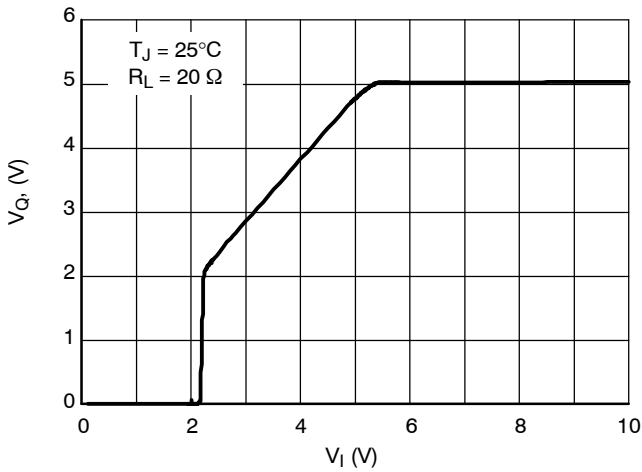


Figure 23. Output Voltage V_Q vs. Input Voltage V_I , 5.0 V Version

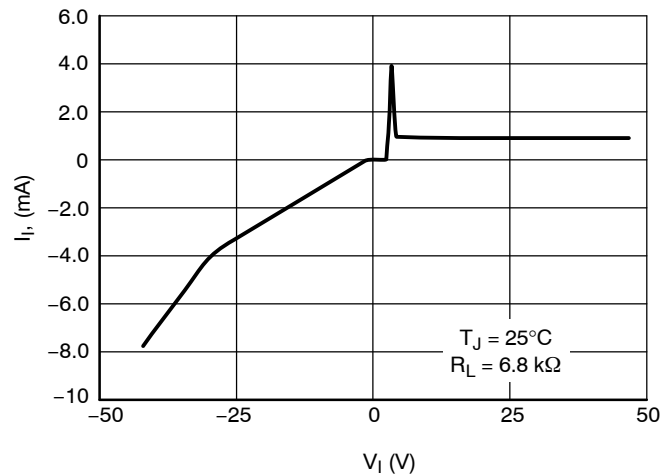


Figure 24. Input Current I_I vs. Input Voltage V_I , 5.0 V Version

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TYPICAL PERFORMANCE CHARACTERISTICS – 4276 Version

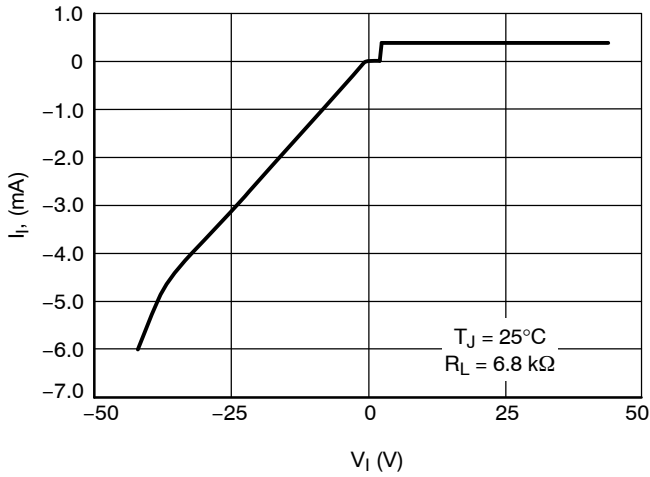


Figure 25. Input Current I_i vs. Input Voltage V_i , 1.8 V Version

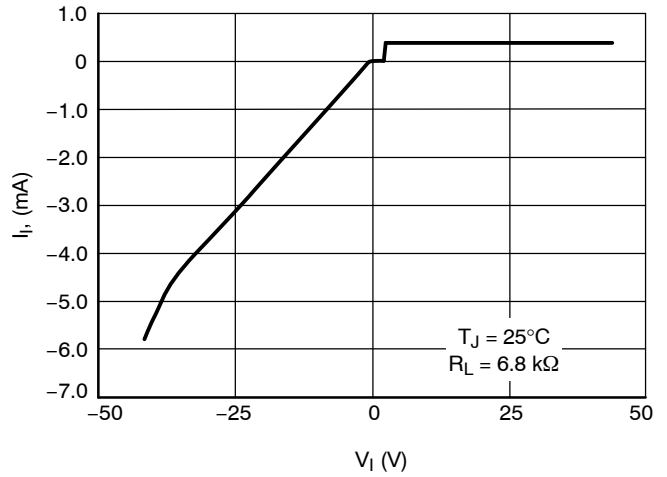


Figure 26. Input Current I_i vs. Input Voltage V_i , 2.5 V Version

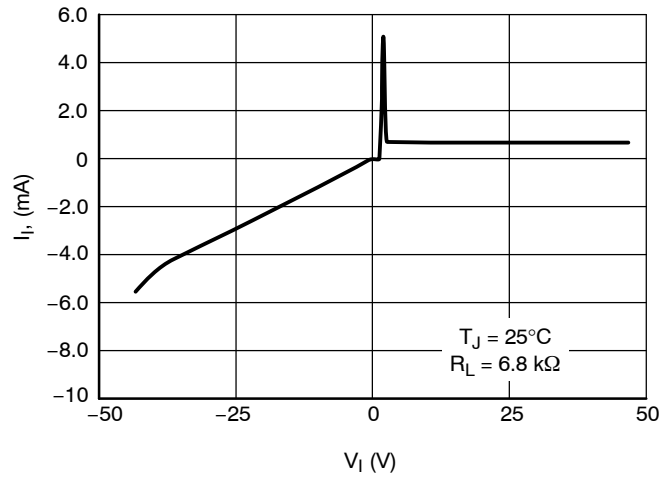


Figure 27. Input Current I_i vs. Input Voltage V_i , 3.3 V Version

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TYPICAL PERFORMANCE CHARACTERISTICS – 4276A Version

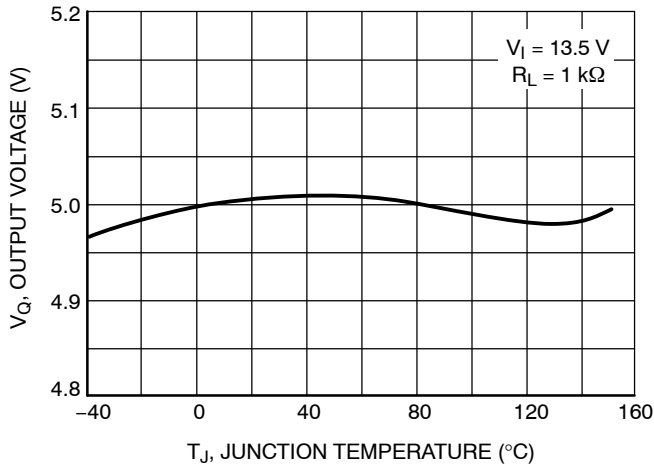


Figure 28. Output Voltage V_Q vs. Junction Temperature T_J , 5.0 V Version

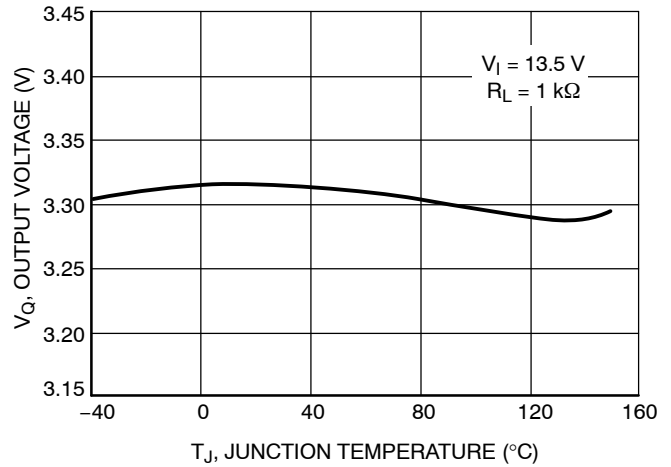


Figure 29. Output Voltage V_Q vs. Junction Temperature T_J , 3.3 V Version

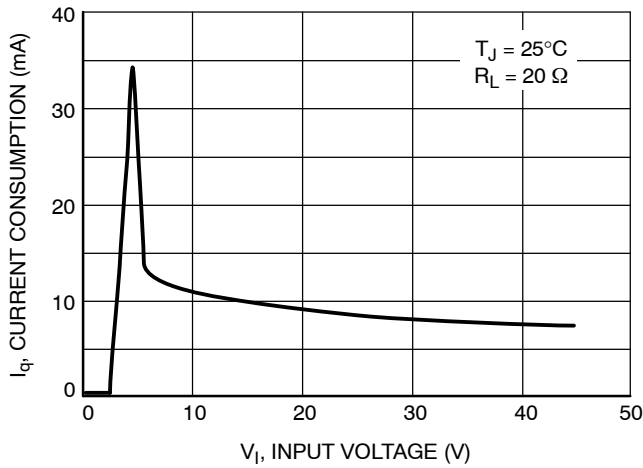


Figure 30. Current Consumption I_q vs. Input Voltage V_I , 5.0 V Version

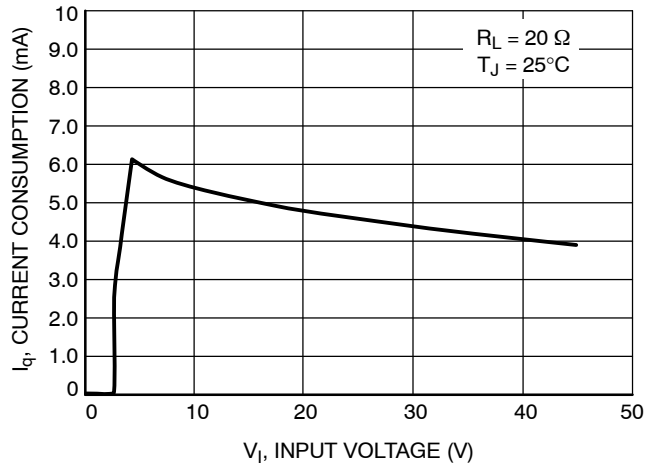


Figure 31. Current Consumption I_q vs. Input Voltage V_I , 3.3 V Version

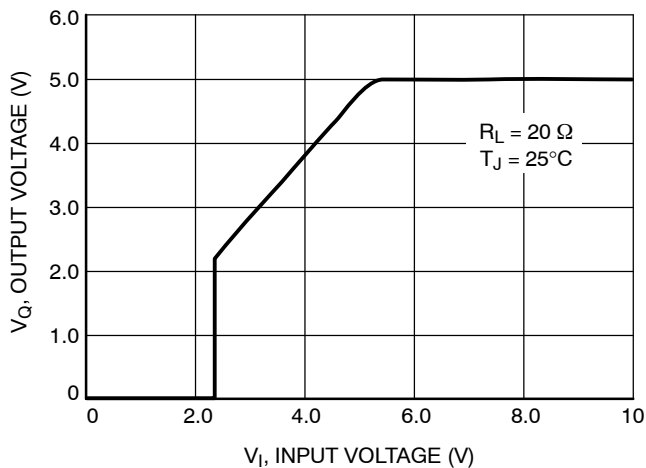


Figure 33. Low Voltage Behavior, 5.0 V Version

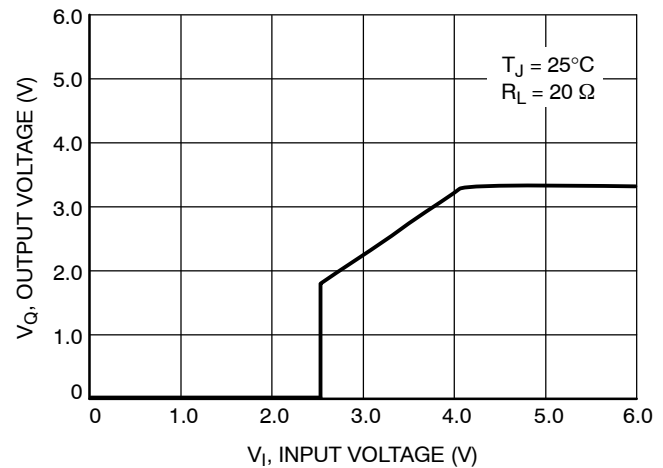


Figure 32. Low Voltage Behavior, 3.3 V Version

TYPICAL PERFORMANCE CHARACTERISTICS – 4276A Version

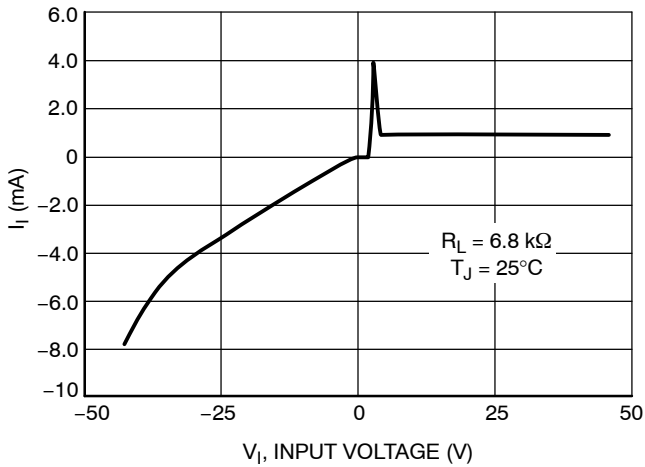


Figure 34. Input Current vs. Input Voltage, 5.0 V Version

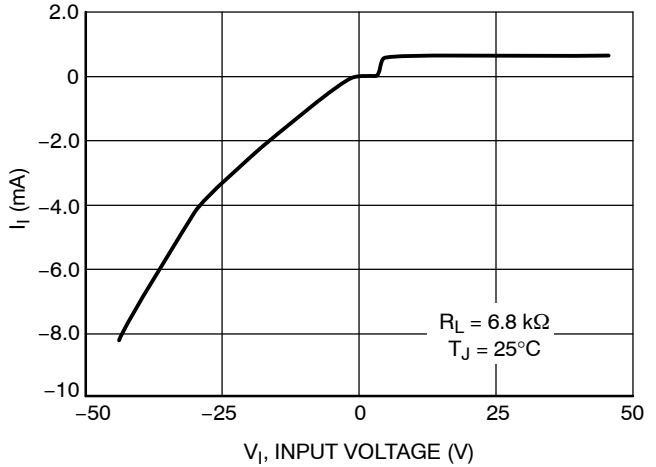


Figure 35. Input Current I_i vs. Input Voltage V_i , 3.3 V Version

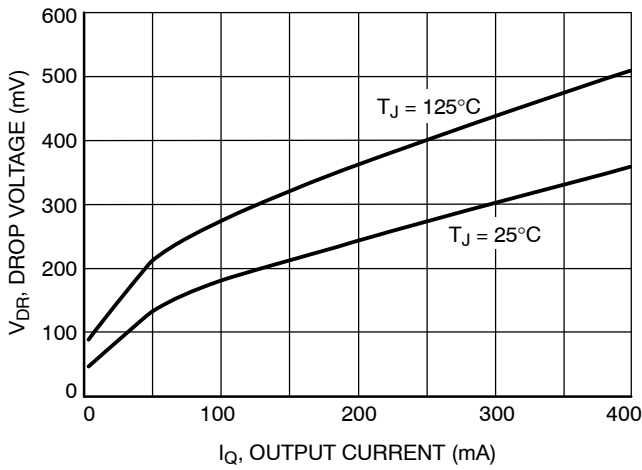


Figure 36. Dropout Voltage V_{DR} vs. Output Current I_Q

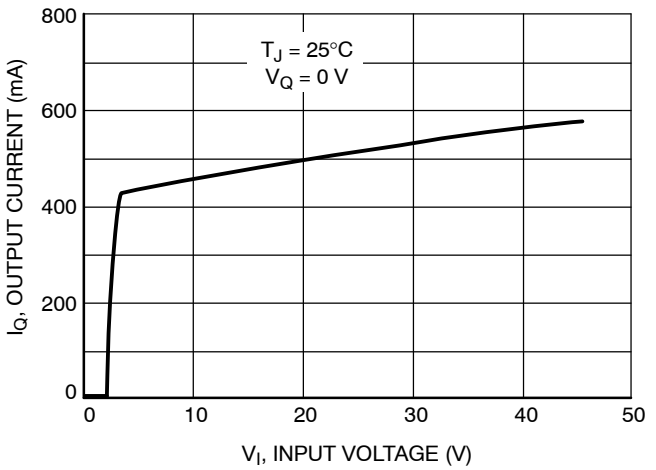


Figure 37. Maximum Output Current I_Q vs. Input Voltage V_i

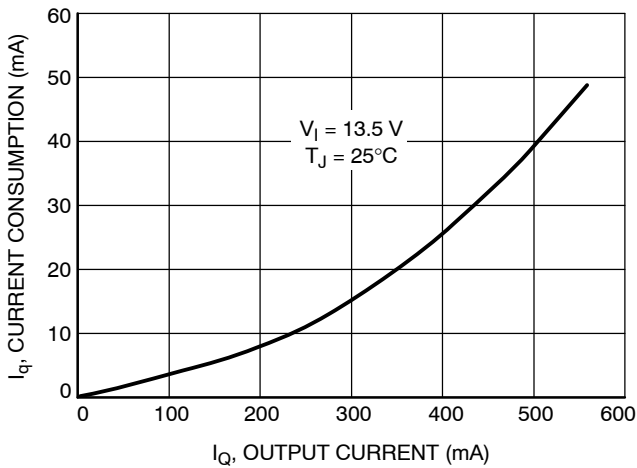


Figure 38. Current Consumption I_q vs. Output Current I_Q (High Load)

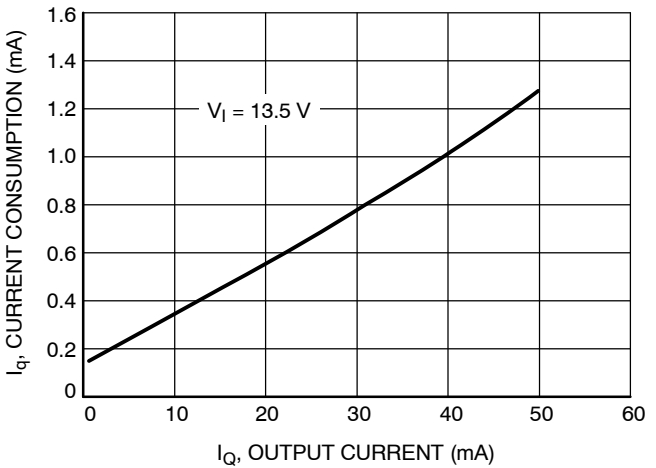


Figure 39. Current Consumption I_q vs. Output Current I_Q (Low Load)

TYPICAL PERFORMANCE CHARACTERISTICS – Adjustable Version

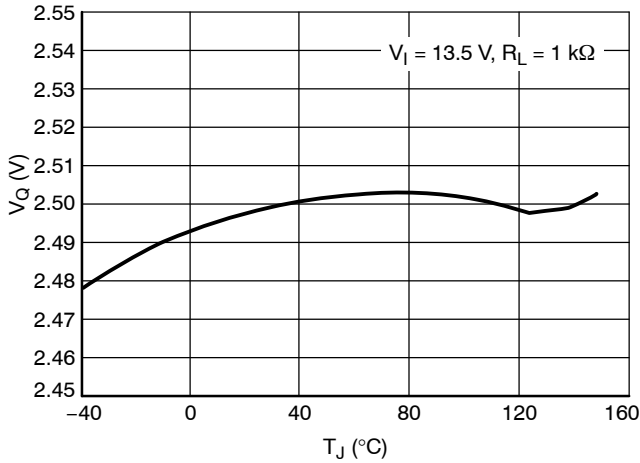


Figure 40. Output Voltage V_Q vs. Junction Temperature T_J , Adjustable Version

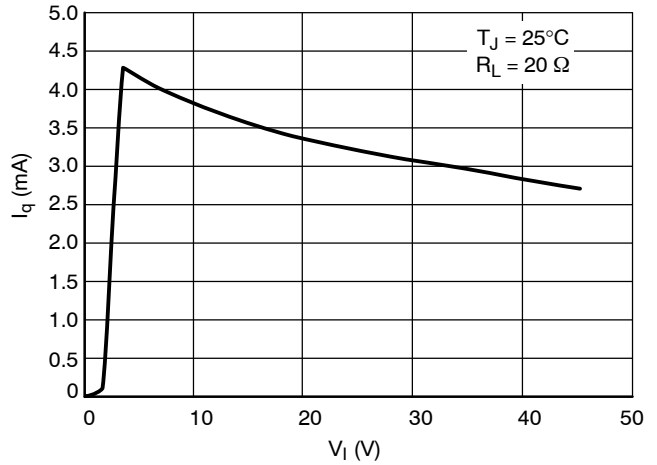


Figure 41. Current Consumption I_q vs. Input Voltage V_I , Adjustable Version

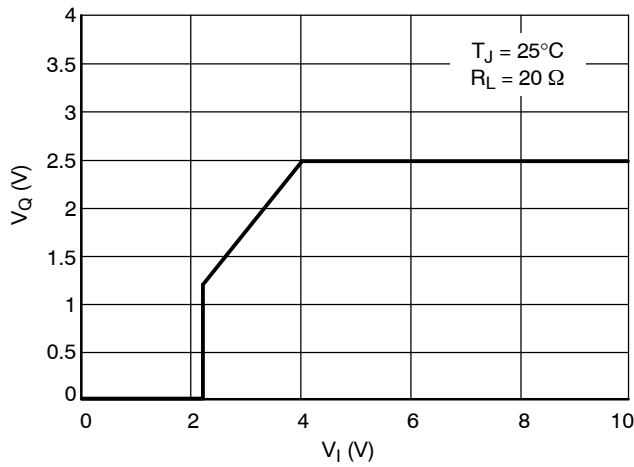


Figure 42. Low Voltage Behavior, Adjustable Version

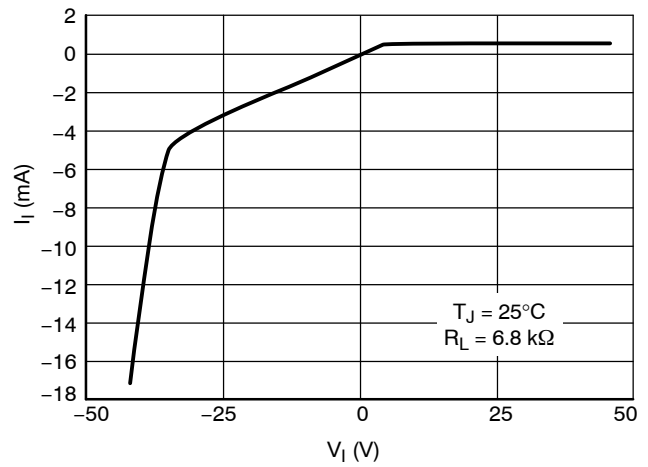


Figure 43. High Voltage Behavior, Adjustable Version

TYPICAL PERFORMANCE CHARACTERISTICS – Adjustable Version

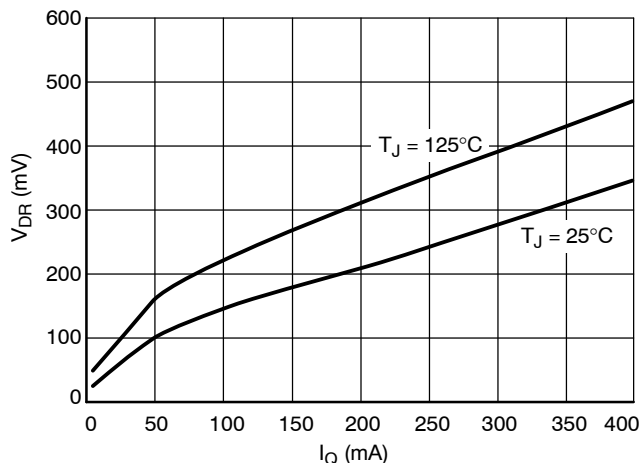


Figure 44. Dropout Voltage V_{DR} vs. Output Current I_Q , Regulator Set at 5.0 V, Adjustable Version

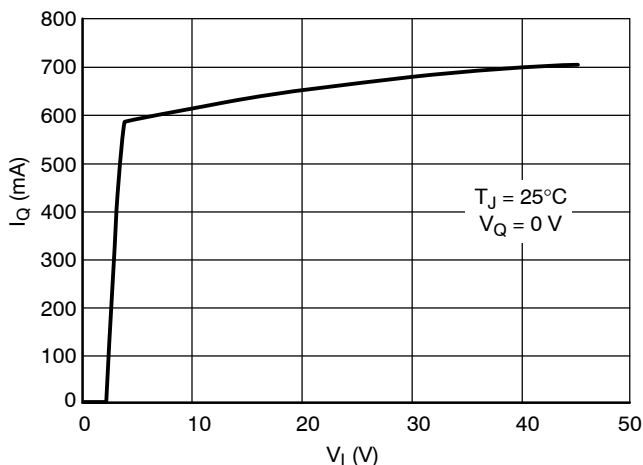


Figure 45. Maximum Output Current I_Q vs. Input Voltage V_I , Adjustable Version

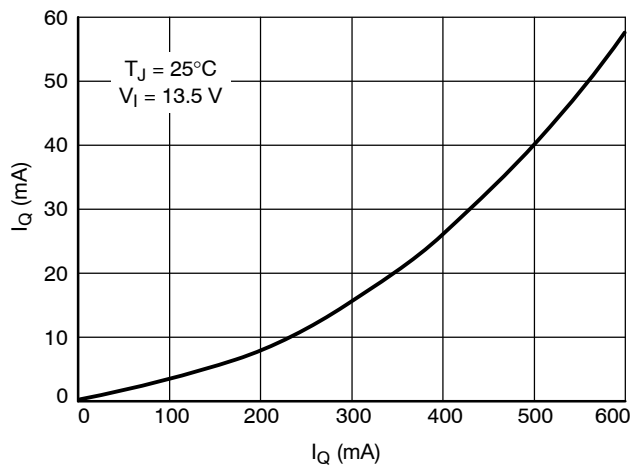


Figure 46. Current Consumption I_q vs. Output Current I_Q (High Load), Adjustable Version

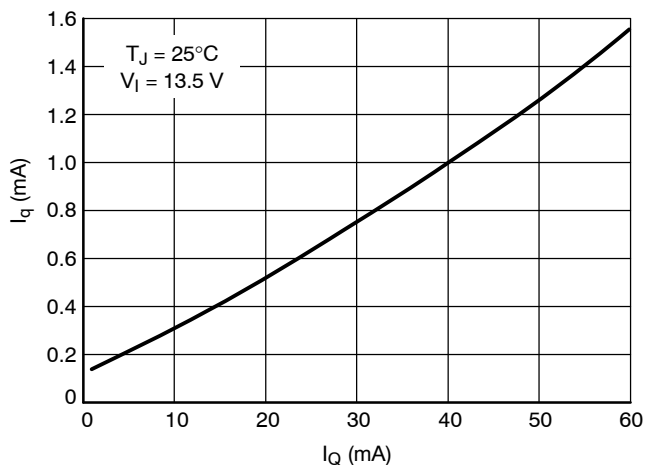


Figure 47. Current Consumption I_q vs. Output Current I_Q (Low Load), Adjustable Version

Circuit Description

The NCV4276 is an integrated low dropout regulator that provides a regulated voltage at 400 mA to the output. It is enabled with an input to the inhibit pin. The regulator voltage is provided by a PNP pass transistor controlled by an error amplifier with a bandgap reference, which gives it the lowest possible dropout voltage. The output current capability is 400 mA, and the base drive quiescent current is controlled to prevent oversaturation when the input voltage is low or when the output is overloaded. The regulator is protected by both current limit and thermal shutdown. Thermal shutdown occurs above 150°C to protect the IC during overloads and extreme ambient temperatures.

Regulator

The error amplifier compares the reference voltage to a sample of the output voltage (V_Q) and drives the base of a PNP series pass transistor via a buffer. The reference is a bandgap design to give it a temperature-stable output. Saturation control of the PNP is a function of the load current and input voltage. Oversaturation of the output power device is prevented, and quiescent current in the ground pin is minimized. See Figure 5, Test Circuit, for circuit element nomenclature illustration.

Regulator Stability Considerations

The input capacitors (C_{I1} and C_{I2}) are necessary to stabilize the input impedance to avoid voltage line influences. Using a resistor of approximately 1.0 Ω in series with C_{I2} can stop potential oscillations caused by stray inductance and capacitance.

The output capacitor helps determine three main characteristics of a linear regulator: startup delay, load transient response and loop stability. The capacitor value and type should be based on cost, availability, size and temperature constraints. The aluminum electrolytic capacitor is the least expensive solution, but, if the circuit operates at low temperatures (-25°C to -40°C), both the value and ESR of the capacitor will vary considerably. The capacitor manufacturer's data sheet usually provides this information.

The value for the output capacitor C_Q , shown in Figure 3, should work for most applications; however, it is not necessarily the optimized solution. Stability is guaranteed for $C_Q \geq 22 \mu\text{F}$ and an $\text{ESR} \leq 2.5 \Omega$ for the 5.0 V and Adjustable regulator and $C_Q \geq 10 \mu\text{F}$ and an $\text{ESR} \leq 1.8 \Omega$ for the 1.8 V, 2.5 V, and 3.3 V regulators. See Figures 5 and 6 for output stability at various load and capacitive ESR conditions.

Inhibit Input

The inhibit pin is used to turn the regulator on or off. By holding the pin down to a voltage less than 0.5 V, the output of the regulator will be turned off. When the voltage on the Inhibit pin is greater than 3.5 V, the output of the regulator will be enabled to power its output to the regulated output voltage. The inhibit pin may be connected directly to the input pin to give constant enable to the output regulator.

Setting the Output Voltage (Adjustable Version)

The output voltage range of the adjustable version can be set between 2.5 V and 20 V (Figure). This is accomplished with an external resistor divider feeding back the voltage to the IC back to the error amplifier by the voltage adjust pin VA. The internal reference voltage is set to a temperature stable reference of 2.5 V.

The output voltage is calculated from the following formula. Ignoring the bias current into the VA pin:

$$V_Q = [(R1 + R2) * V_{ref}] / R2$$

Use $R2 < 50 \text{ k}$ to avoid significant voltage output errors due to VA bias current.

Connecting VA directly to Q without R1 and R2 creates an output voltage of 2.5 V.

Designers should consider the tolerance of R1 and R2 during the design phase.

The input voltage range for operation (pin 1) of the adjustable version is between ($V_Q + 0.5 \text{ V}$) and 40 V. Internal bias requirements dictate a minimum input voltage of 4.5 V. The dropout voltage for output voltages less than 4.0 V is ($4.5 \text{ V} - V_Q$).

Calculating Power Dissipation in a Single Output Linear Regulator

The maximum power dissipation for a single output regulator (Figure 48) is:

$$P_{D(max)} = [V_{I(max)} - V_{Q(min)}] I_{Q(max)} + V_{I(max)} I_q \quad (1)$$

where

- $V_{I(max)}$ is the maximum input voltage,
- $V_{Q(min)}$ is the minimum output voltage,
- $I_{Q(max)}$ is the maximum output current for the application,
- I_q is the quiescent current the regulator consumes at $I_{Q(max)}$.

Once the value of $P_{D(max)}$ is known, the maximum permissible value of $R_{\theta JA}$ can be calculated:

$$R_{\theta JA} = \frac{150^{\circ}\text{C} - T_A}{P_D} \quad (2)$$

The value of $R_{\theta JA}$ can then be compared with those in the package section of the data sheet. Those packages with $R_{\theta JA}$ less than the calculated value in Equation 2 will keep the die temperature below 150°C.

In some cases, none of the packages will be sufficient to dissipate the heat generated by the IC, and an external heatsink will be required.

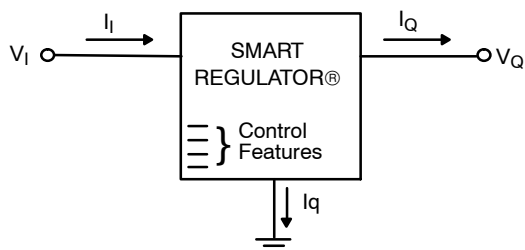


Figure 48. Single Output Regulator with Key Performance Parameters Labeled

Heatsinks

A heatsink effectively increases the surface area of the package to improve the flow of heat away from the IC and into the surrounding air.

Each material in the heat flow path between the IC and the outside environment will have a thermal resistance. Like series electrical resistances, these resistances are summed to determine the value of $R_{\theta JA}$:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CS} + R_{\theta SA} \quad (3)$$

where

- $R_{\theta JC}$ is the junction-to-case thermal resistance,
- $R_{\theta CS}$ is the case-to-heatsink thermal resistance,
- $R_{\theta SA}$ is the heatsink-to-ambient thermal resistance.

$R_{\theta JC}$ appears in the package section of the data sheet. Like $R_{\theta JA}$, it too is a function of package type. $R_{\theta CS}$ and $R_{\theta SA}$ are functions of the package type, heatsink and the interface between them. These values appear in data sheets of heatsink manufacturers.

Thermal, mounting, and heatsinking considerations are discussed in the ON Semiconductor application note AN1040/D.

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Thermal Model

A discussion of thermal modeling is in the ON Semiconductor web site: <http://www.onsemi.com/pub/collateral/BR1487-D.PDF>.

Table 1. DPAK 5-Lead Thermal RC Network Models

Drain Copper Area (1 oz thick)			168 mm ²	736 mm ²		168 mm ²	736 mm ²	
(SPICE Deck Format)			Cauer Network			Foster Network		
			168 mm ²	736 mm ²	Units	Tau	Tau	Units
C_C1	Junction	GND	1.00E-06	1.00E-06	W-s/C	1.36E-08	1.361E-08	sec
C_C2	node1	GND	1.00E-05	1.00E-05	W-s/C	7.41E-07	7.411E-07	sec
C_C3	node2	GND	6.00E-05	6.00E-05	W-s/C	1.04E-05	1.029E-05	sec
C_C4	node3	GND	1.00E-04	1.00E-04	W-s/C	3.91E-05	3.737E-05	sec
C_C5	node4	GND	4.36E-04	3.64E-04	W-s/C	1.80E-03	1.376E-03	sec
C_C6	node5	GND	6.77E-02	1.92E-02	W-s/C	3.77E-01	2.851E-02	sec
C_C7	node6	GND	1.51E-01	1.27E-01	W-s/C	3.79E+00	9.475E-01	sec
C_C8	node7	GND	4.80E-01	1.018	W-s/C	2.65E+01	1.173E+01	sec
C_C9	node8	GND	3.740	2.955	W-s/C	8.71E+01	8.59E+01	sec
C_C10	node9	GND	10.322	0.438	W-s/C			sec
			168 mm ²	736 mm ²		R's	R's	
R_R1	Junction	node1	0.015	0.015	C/W	0.0123	0.0123	C/W
R_R2	node1	node2	0.08	0.08	C/W	0.0585	0.0585	C/W
R_R3	node2	node3	0.4	0.4	C/W	0.0304	0.0287	C/W
R_R4	node3	node4	0.2	0.2	C/W	0.3997	0.3772	C/W
R_R5	node4	node5	2.97519	2.6171	C/W	3.115	2.68	C/W
R_R6	node5	node6	8.2971	1.6778	C/W	3.571	1.38	C/W
R_R7	node6	node7	25.9805	7.4246	C/W	12.851	5.92	C/W
R_R8	node7	node8	46.5192	14.9320	C/W	35.471	7.39	C/W
R_R9	node8	node9	17.7808	19.2560	C/W	46.741	28.94	C/W
R_R10	node9	GND	0.1	0.1758	C/W			C/W

NOTE: Bold face items represent the package without the external thermal system.

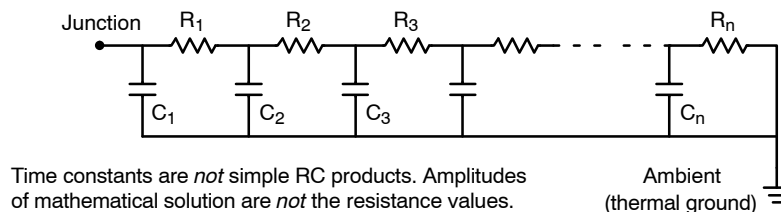


Figure 49. Grounded Capacitor Thermal Network ("Cauer" Ladder)

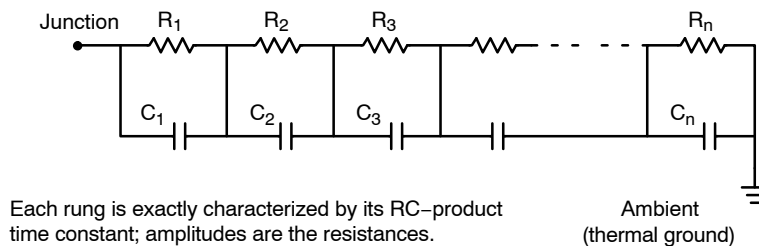


Figure 50. Non-Grounded Capacitor Thermal Ladder ("Foster" Ladder)

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Table 2. D²PAK 5-Lead Thermal RC Network Models

Drain Copper Area (1 oz thick)			241 mm ²		788 mm ²		241 mm ²			788 mm ²		
			(SPICE Deck Format)			Cauer Network			Foster Network			
			241 mm ²	653 mm ²	Units	Tau	Tau	Units				
C_C1	Junction	GND	1.00E-06	1.00E-06	W-s/C	1.361E-08	1.361E-08	sec				
C_C2	node1	GND	1.00E-05	1.00E-05	W-s/C	7.411E-07	7.411E-07	sec				
C_C3	node2	GND	6.00E-05	6.00E-05	W-s/C	1.005E-05	1.007E-05	sec				
C_C4	node3	GND	1.00E-04	1.00E-04	W-s/C	3.460E-05	3.480E-05	sec				
C_C5	node4	GND	2.82E-04	2.87E-04	W-s/C	7.868E-04	8.107E-04	sec				
C_C6	node5	GND	5.58E-03	5.95E-03	W-s/C	7.431E-03	7.830E-03	sec				
C_C7	node6	GND	4.25E-01	4.61E-01	W-s/C	2.786E+00	2.012E+00	sec				
C_C8	node7	GND	9.22E-01	2.05	W-s/C	2.014E+01	2.601E+01	sec				
C_C9	node8	GND	1.73	4.88	W-s/C	1.134E+02	1.218E+02	sec				
C_C10	node9	GND	7.12	1.31	W-s/C			sec				
			241 mm ²	653 mm ²		R's	R's					
R_R1	Junction	node1	0.015	0.0150	C/W	0.0123	0.0123	C/W				
R_R2	node1	node2	0.08	0.0800	C/W	0.0585	0.0585	C/W				
R_R3	node2	node3	0.4	0.4000	C/W	0.0257	0.0260	C/W				
R_R4	node3	node4	0.2	0.2000	C/W	0.3413	0.3438	C/W				
R_R5	node4	node5	1.85638	1.8839	C/W	1.77	1.81	C/W				
R_R6	node5	node6	1.23672	1.2272	C/W	1.54	1.52	C/W				
R_R7	node6	node7	9.81541	5.3383	C/W	4.13	3.46	C/W				
R_R8	node7	node8	33.1868	18.9591	C/W	6.27	5.03	C/W				
R_R9	node8	node9	27.0263	13.3369	C/W	60.80	29.30	C/W				
R_R10	node9	GND	1.13944	0.1191	C/W			C/W				

NOTE: Bold face items represent the package without the external thermal system.

The Cauer networks generally have physical significance and may be divided between nodes to separate thermal behavior due to one portion of the network from another. The Foster networks, though when sorted by time constant (as above) bear a rough correlation with the Cauer networks, are really only convenient mathematical models. Cauer networks can be easily implemented using circuit simulating tools, whereas Foster networks may be more easily implemented using mathematical tools (for instance, in a spreadsheet program), according to the following formula:

$$R(t) = \sum_{i=1}^n R_i (1 - e^{-t/\tau_i})$$

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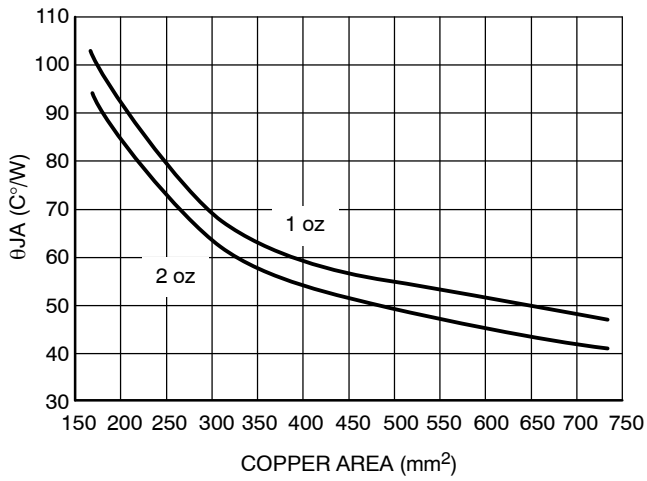


Figure 51. θ_{JA} vs. Copper Spreader Area, DPAK 5-Lead

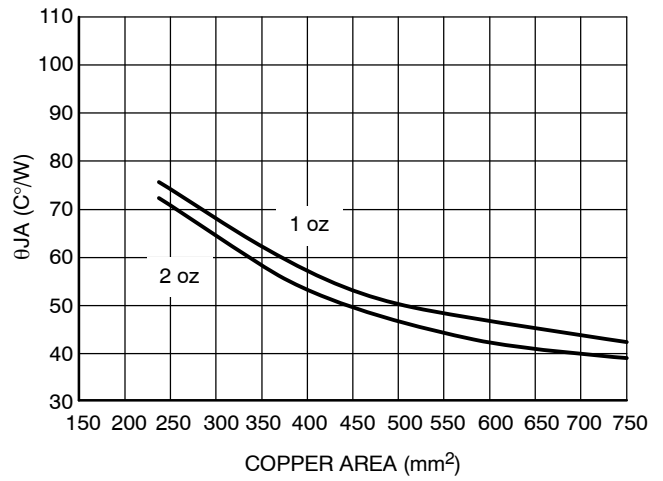


Figure 52. θ_{JA} vs. Copper Spreader Area, D²PAK 5-Lead

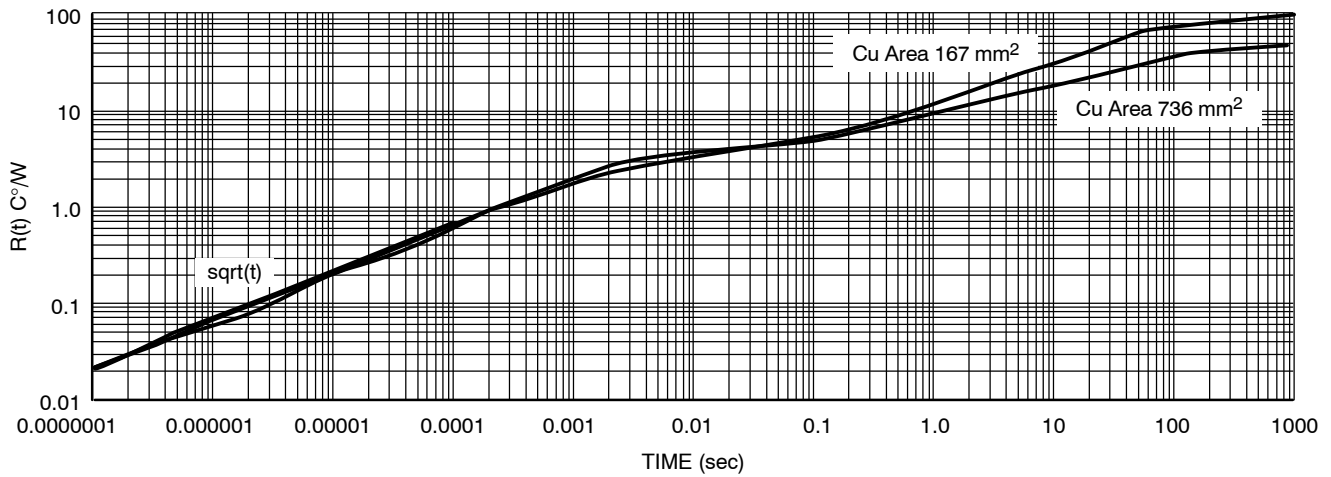


Figure 53. Single-Pulse Heating Curves, DPAK 5-Lead

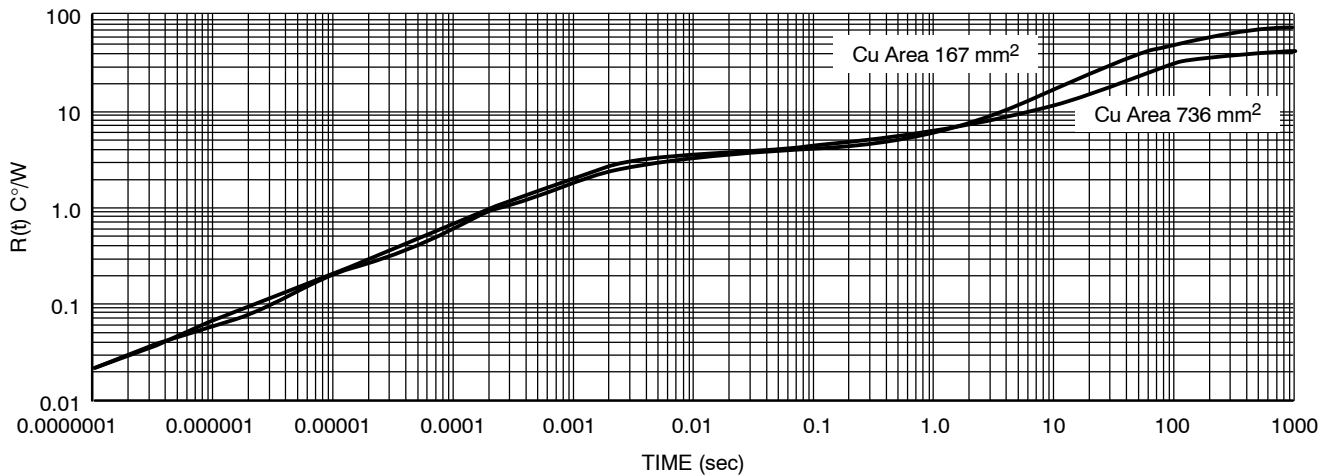


Figure 54. Single-Pulse Heating Curves, D²PAK 5-Lead

NCV4276, NCV4276A

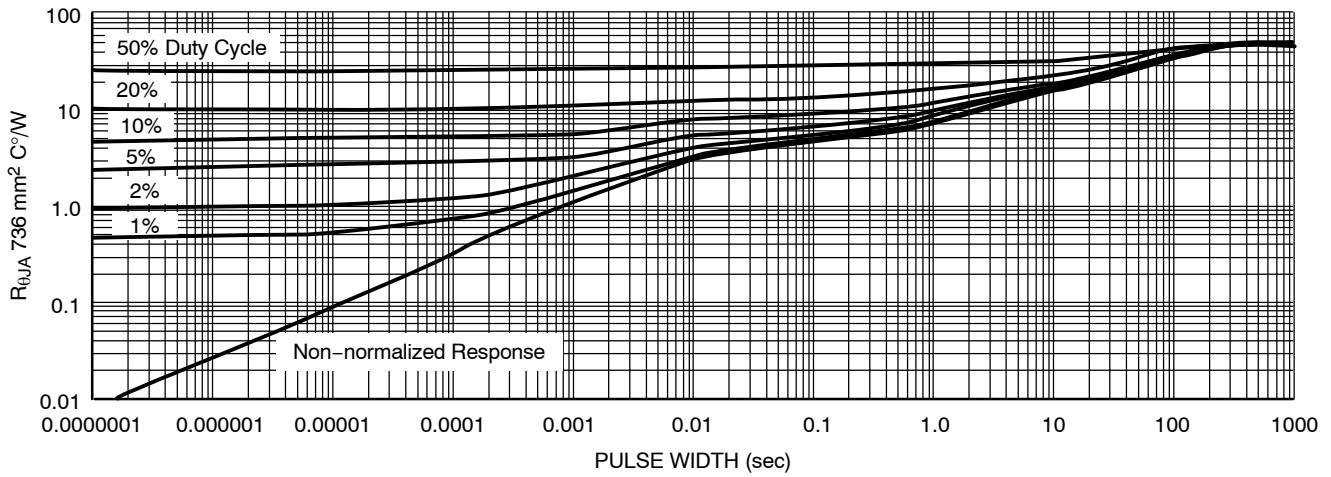


Figure 55. Duty Cycle for 1" Spreader Boards, DPAK 5-Lead

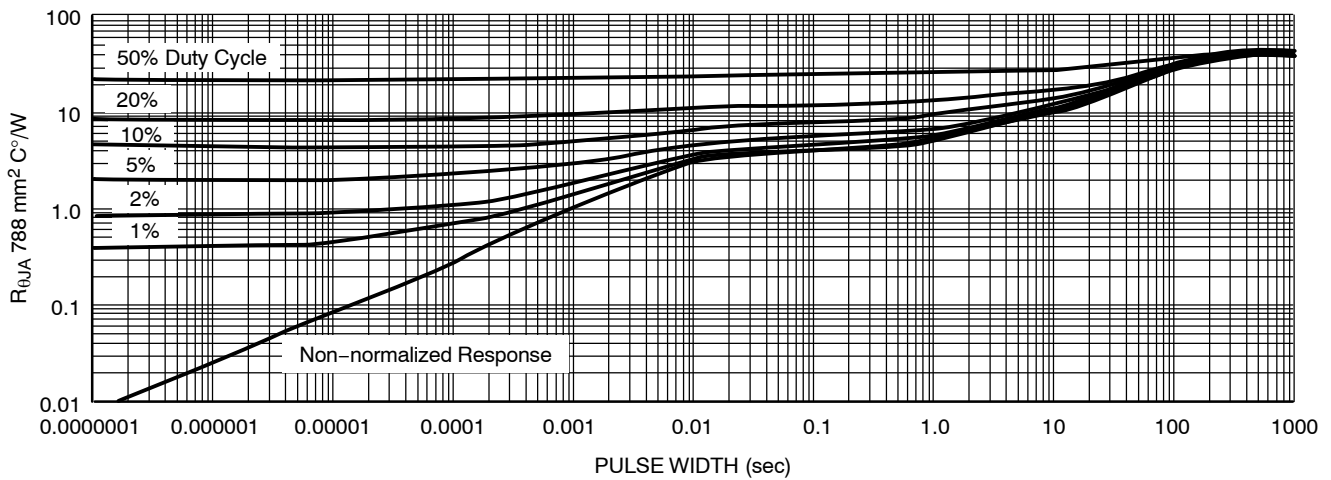
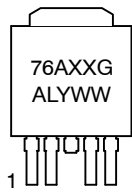


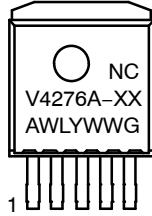
Figure 56. Duty Cycle for 1" Spreader Boards, D²PAK 5-Lead

NCV4276, NCV4276A

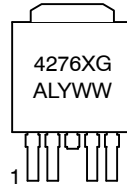
MARKING DIAGRAMS



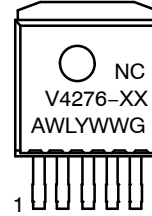
NCV4276A
DPAK
5-PIN
DT SUFFIX
CASE 175AA



NCV4276A
D²PAK
5-PIN
DS SUFFIX
CASE 936A



NCV4276
DPAK
5-PIN
DT SUFFIX
CASE 175AA



NCV4276
D²PAK
5-PIN
DS SUFFIX
CASE 936A

*Tab is connected to Pin 3 on all packages.

A = Assembly Location
 WL, L = Wafer Lot
 Y = Year
 WW = Work Week
 G = Pb-Free Device
 x, xx = Voltage Ratings as indicated below

A-Version

DPAK	D ² PAK
XX = AJ (Adj. Voltage)	XX = AJ (Adj. Voltage)
XX = 50 (5.0 V)	XX = 50 (5.0 V)
XX = 33 (3.3 V)	

Non-A-Version

DPAK	D ² PAK
X = V (Adj. Voltage)	XX = AJ (Adj. Voltage)
X = 5 (5.0 V)	XX = 50 (5.0 V)
X = 3 (3.3 V)	XX = 33 (3.3 V)
	XX = 25 (2.5 V)
	XX = 18 (1.8 V)

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ORDERING INFORMATION

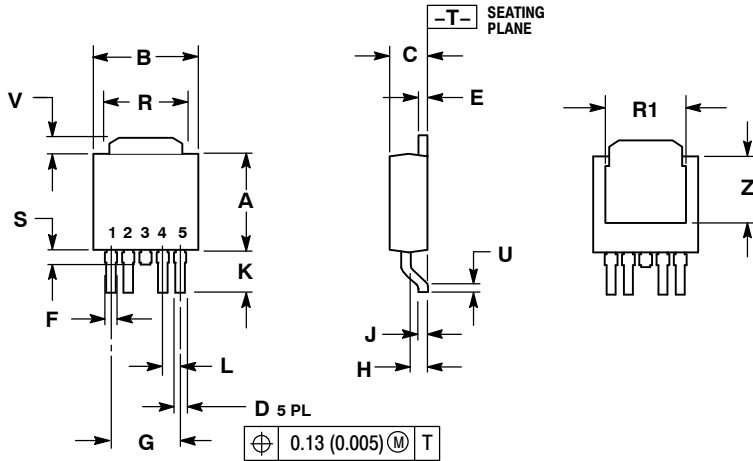
Device	Output Voltage Accuracy	Output Voltage	Package	Shipping†
NCV4276DT50RK	4%	5.0 V	DPAK, 5-Pin	2500 / Tape & Reel
NCV4276DT50RKG			DPAK, 5-Pin (Pb-Free)	2500 / Tape & Reel
NCV4276DS50			D ² PAK, 5-Pin	50 Units / Rail
NCV4276DS50G			D ² PAK, 5-Pin (Pb-Free)	50 Units / Rail
NCV4276DS50R4			D ² PAK, 5-Pin	800 / Tape & Reel
NCV4276DS50R4G			D ² PAK, 5-Pin (Pb-Free)	800 / Tape & Reel
NCV4276DT33RK		3.3 V	DPAK, 5-Pin	2500 / Tape & Reel
NCV4276DT33RKG			DPAK, 5-Pin (Pb-Free)	2500 / Tape & Reel
NCV4276DS33			D ² PAK, 5-Pin	50 Units / Rail
NCV4276DS33G			D ² PAK, 5-Pin (Pb-Free)	50 Units / Rail
NCV4276DS33R4			D ² PAK, 5-Pin	800 / Tape & Reel
NCV4276DS33R4G			D ² PAK, 5-Pin (Pb-Free)	800 / Tape & Reel
NCV4276DS25		2.5 V	D ² PAK, 5-Pin	50 Units / Rail
NCV4276DS25G			D ² PAK, 5-Pin (Pb-Free)	50 Units / Rail
NCV4276DS25R4			D ² PAK, 5-Pin	800 / Tape & Reel
NCV4276DS25R4G			D ² PAK, 5-Pin (Pb-Free)	800 / Tape & Reel
NCV4276DS18		1.8 V	D ² PAK, 5-Pin	50 Units / Rail
NCV4276DS18G			D ² PAK, 5-Pin (Pb-Free)	50 Units / Rail
NCV4276DS18R4			D ² PAK, 5-Pin	800 / Tape & Reel
NCV4276DS18R4G			D ² PAK, 5-Pin (Pb-Free)	800 / Tape & Reel
NCV4276DTADJRKG		Adjustable	DPAK, 5-Pin (Pb-Free)	2500 / Tape & Reel
NCV4276DSADJG			D ² PAK, 5-Pin (Pb-Free)	50 Units / Rail
NCV4276DSADJR4G			D ² PAK, 5-Pin (Pb-Free)	800 / Tape & Reel
NCV4276ADT33RKG		2%	3.3 V	DPAK, 5-Pin (Pb-Free)
NCV4276ADT50RKG	5.0 V		DPAK, 5-Pin (Pb-Free)	2500 / Tape & Reel
NCV4276ADS50G			D ² PAK, 5-Pin (Pb-Free)	50 Units / Rail
NCV4276ADS50R4G	D ² PAK, 5-Pin (Pb-Free)		800 / Tape & Reel	
NCV4276ADTADJRKG	Adjustable		DPAK, 5-Pin (Pb-Free)	2500 / Tape & Reel
NCV4276ADSADJG			D ² PAK, 5-Pin (Pb-Free)	50 Units / Rail
NCV4276ADSADJR4G			D ² PAK, 5-Pin (Pb-Free)	800 / Tape & Reel
NCV4276ADSADJR4G			D ² PAK, 5-Pin (Pb-Free)	800 / Tape & Reel

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

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PACKAGE DIMENSIONS

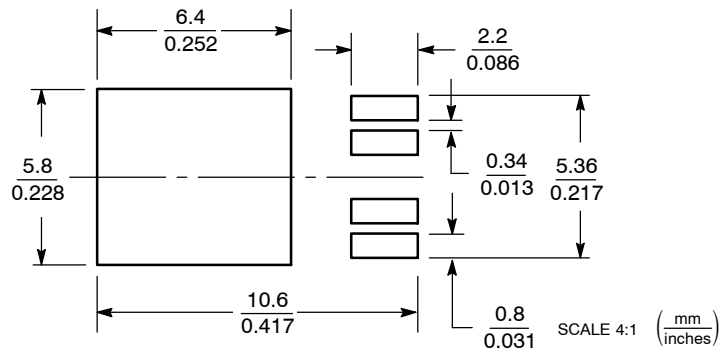
DPAK 5, CENTER LEAD CROP
DT SUFFIX
CASE 175AA-01
ISSUE A



- NOTES:
1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: INCH.

DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	0.235	0.245	5.97	6.22
B	0.250	0.265	6.35	6.73
C	0.086	0.094	2.19	2.38
D	0.020	0.028	0.51	0.71
E	0.018	0.023	0.46	0.58
F	0.024	0.032	0.61	0.81
G	0.180	BSC	4.56	BSC
H	0.034	0.040	0.87	1.01
J	0.018	0.023	0.46	0.58
K	0.102	0.114	2.60	2.89
L	0.045	BSC	1.14	BSC
R	0.170	0.190	4.32	4.83
R1	0.185	0.210	4.70	5.33
S	0.025	0.040	0.63	1.01
U	0.020	---	0.51	---
V	0.035	0.050	0.89	1.27
Z	0.155	0.170	3.93	4.32

SOLDERING FOOTPRINT*

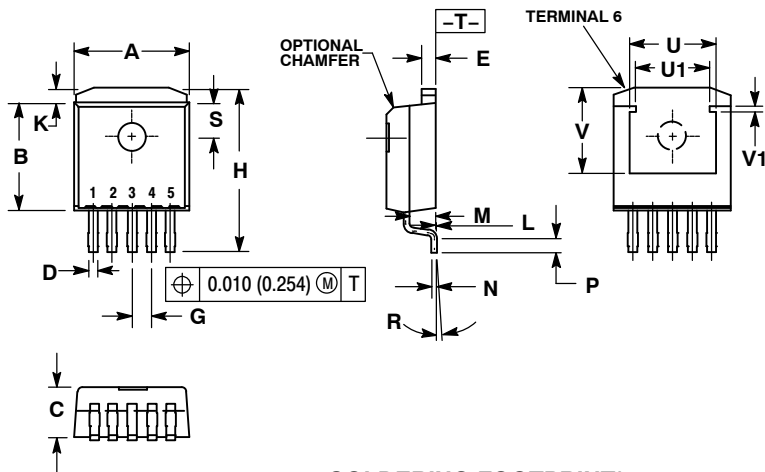


*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

NCV4276, NCV4276A

PACKAGE DIMENSIONS

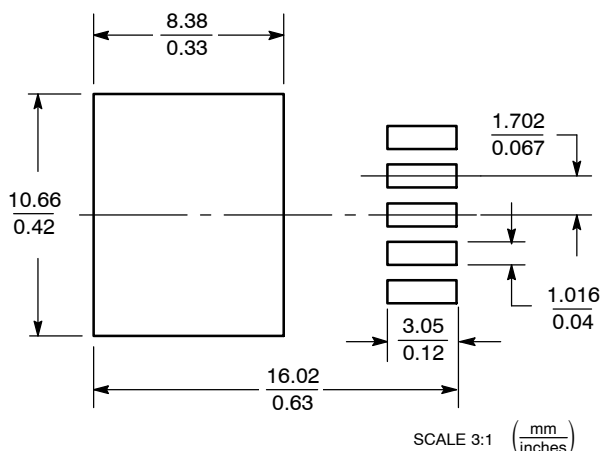
D²PAK 5 CASE 936A-02 ISSUE D



- NOTES:
1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
 2. CONTROLLING DIMENSION: INCH.
 3. TAB CONTOUR OPTIONAL WITHIN DIMENSIONS A AND K.
 4. DIMENSIONS U AND V ESTABLISH A MINIMUM MOUNTING SURFACE FOR TERMINAL 6.
 5. DIMENSIONS A AND B DO NOT INCLUDE MOLD FLASH OR GATE PROTRUSIONS. MOLD FLASH AND GATE PROTRUSIONS NOT TO EXCEED 0.025 (0.635) MAXIMUM.

DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	0.386	0.403	9.804	10.236
B	0.356	0.368	9.042	9.347
C	0.170	0.180	4.318	4.572
D	0.026	0.036	0.660	0.914
E	0.045	0.055	1.143	1.397
G	0.067 BSC		1.702 BSC	
H	0.539	0.579	13.691	14.707
K	0.050 REF		1.270 REF	
L	0.000	0.010	0.000	0.254
M	0.088	0.102	2.235	2.591
N	0.018	0.026	0.457	0.660
P	0.058	0.078	1.473	1.981
R	5° REF		5° REF	
S	0.116 REF		2.946 REF	
U	0.200 MIN		5.080 MIN	
V	0.250 MIN		6.350 MIN	
U1	0.297	0.305	7.544	7.747
V1	0.038	0.046	0.965	1.168

SOLDERING FOOTPRINT*



SCALE 3:1 (mm/inches)

*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

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